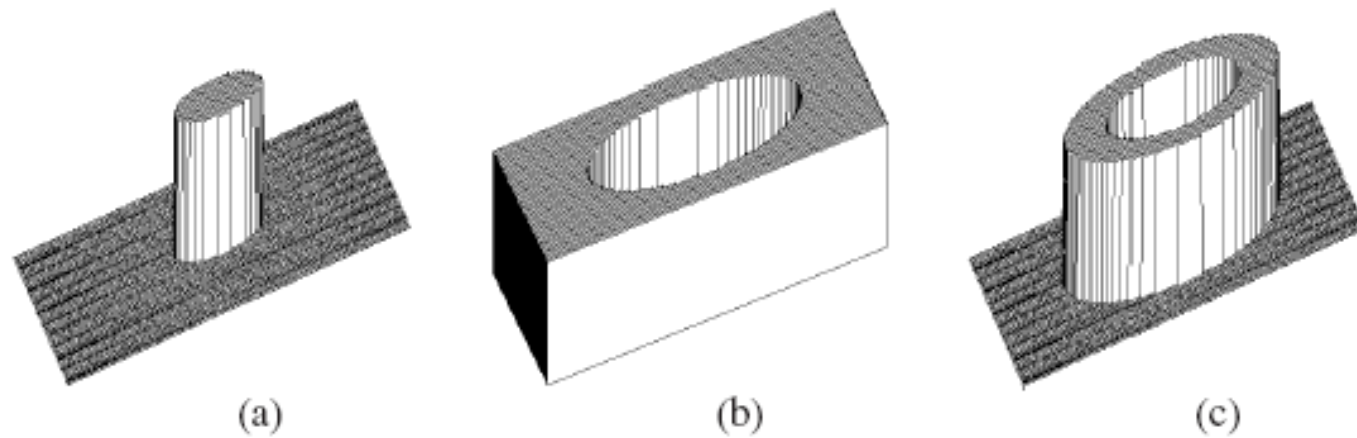


# The Discrete Time Fourier Transform-4. Properties (cont'd)

# Today

- Applications of the Fourier Transform in image analysis
- Properties of the Fourier Transform (cont'd)



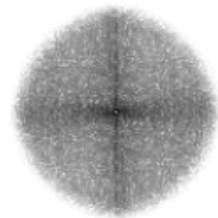
**Figure 5.26:** Frequency filters displayed in 3D. (a) Low-pass filter. (b) High-pass filter. (c) Band-pass filter.



(a)



(b)

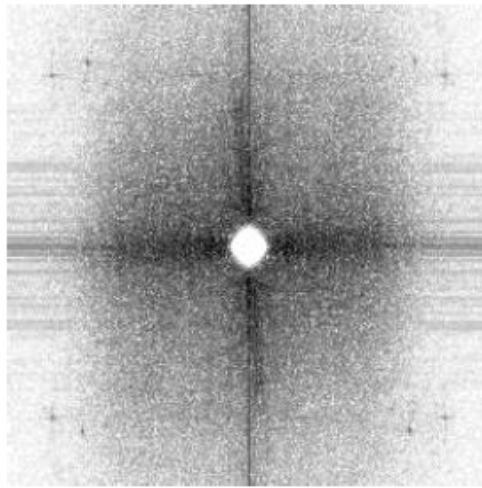


(c)



(d)

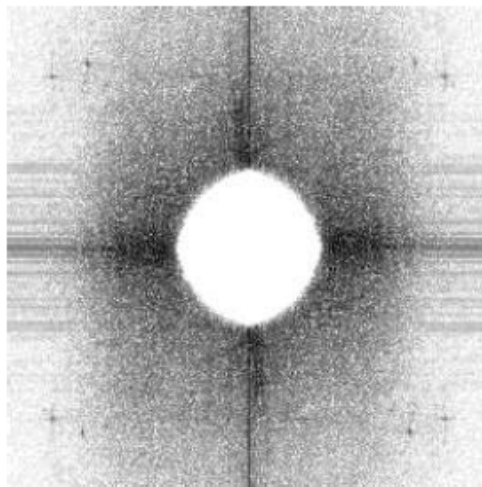
**Figure 5.27:** Low-pass frequency-domain filtering—for the original image and its spectrum see Figure 3.7. (a) Spectrum of a low-pass filtered image, all higher frequencies filtered out. (b) Image resulting from the inverse Fourier transform applied to spectrum (a). (c) Spectrum of a low-pass filtered image, only very high frequencies filtered out. (d) Inverse Fourier transform applied to spectrum (c).



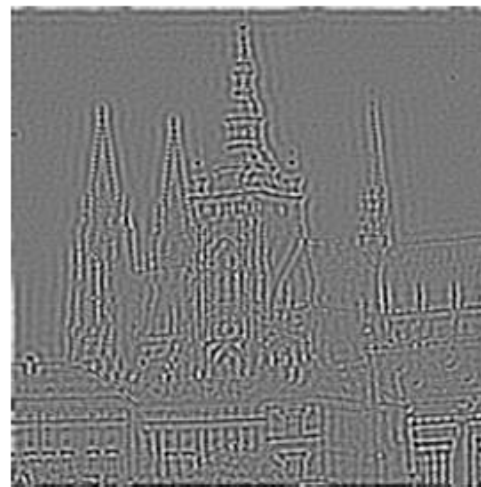
(a)



(b)



(c)

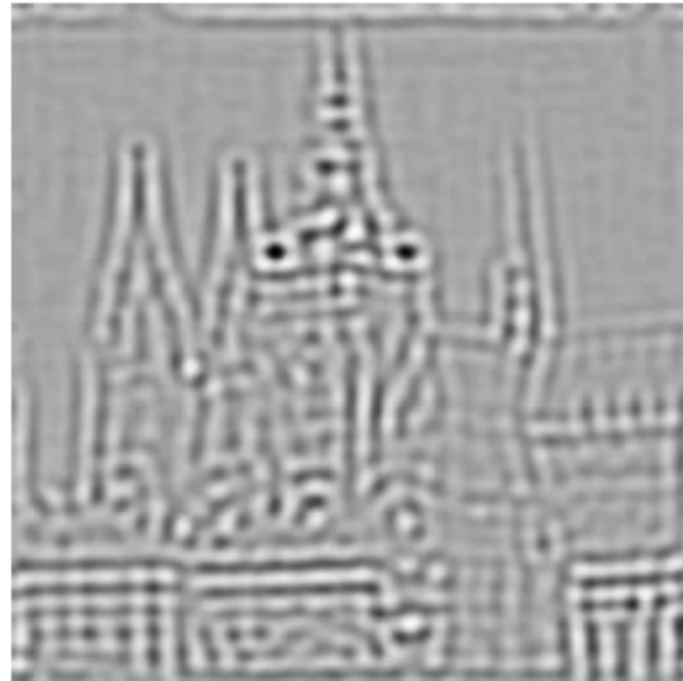


(d)

**Figure 5.28:** High-pass frequency domain filtering. (a) Spectrum of a high-pass filtered image, only very low frequencies filtered out. (b) Image resulting from the inverse Fourier transform applied to spectrum (a). (c) Spectrum of a high-pass filtered image, all lower frequencies filtered out. (d) Inverse Fourier transform applied to spectrum (c).



(a)

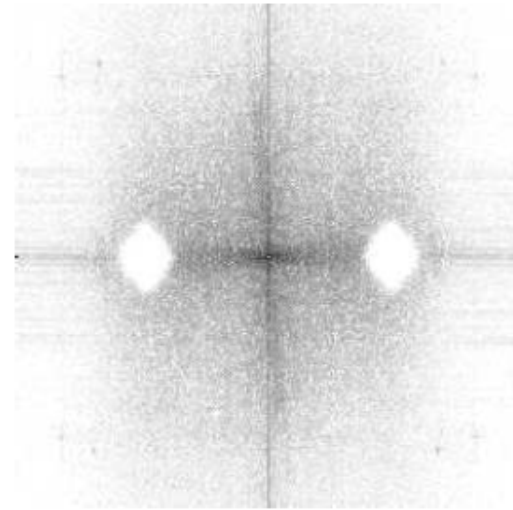


(b)

**Figure 5.29:** Band-pass frequency domain filtering. (a) Spectrum of a band-pass-filtered image, low and high frequencies filtered out. (b) Image resulting from the inverse Fourier transform applied to spectrum (a).



(a)



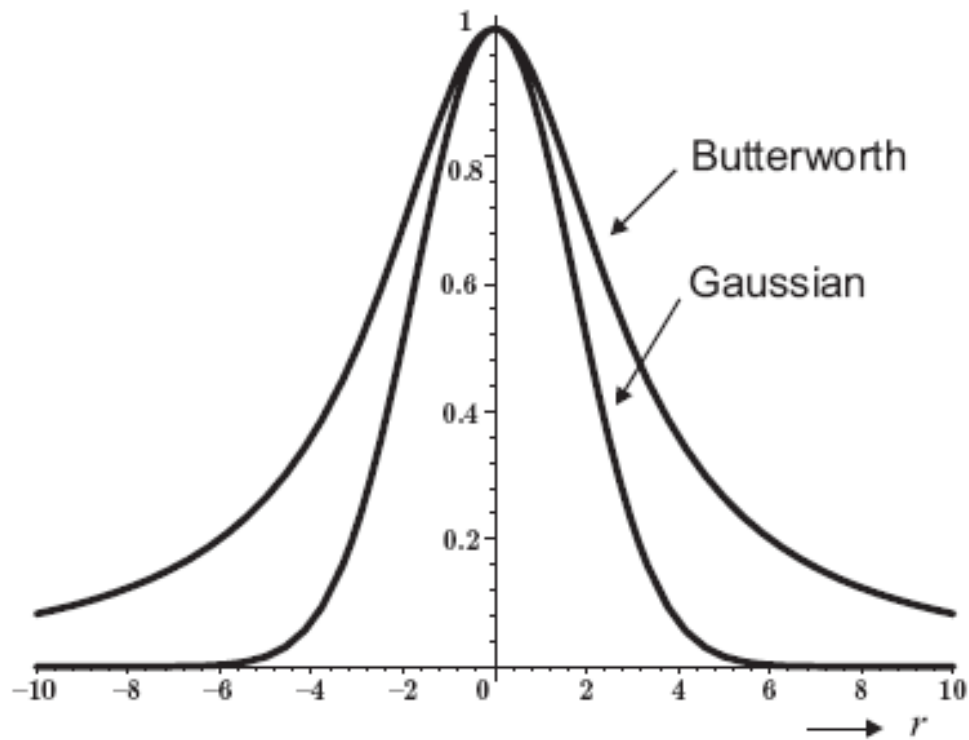
(b)



(c)

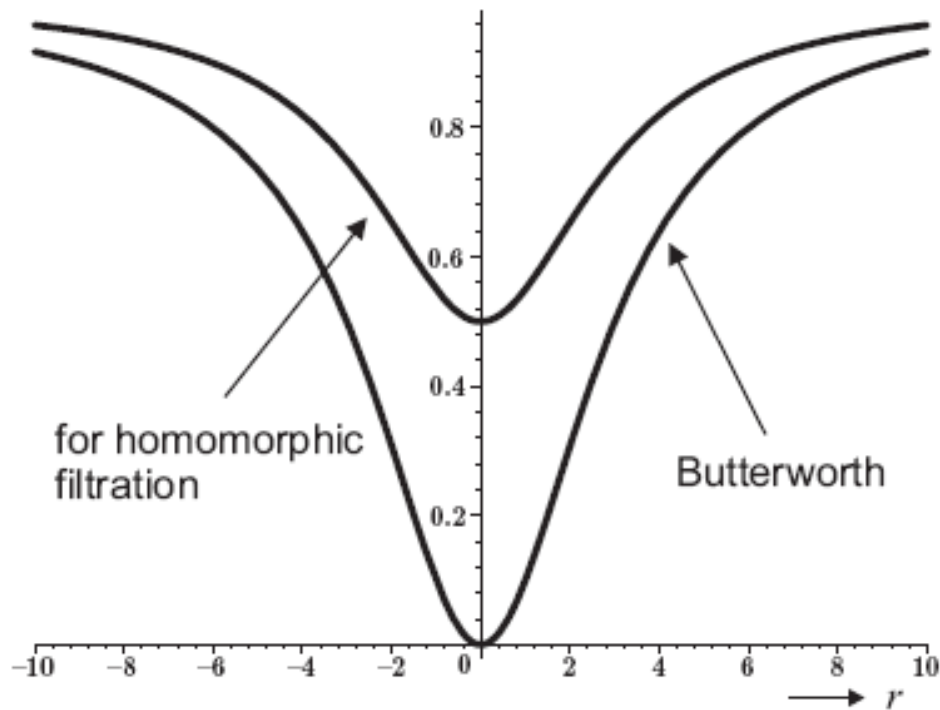
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**Figure 5.30:** Periodic noise removal. (a) Noisy image. (b) Image spectrum used for image reconstruction—note that the areas of frequencies corresponding with periodic vertical lines are filtered out. (c) Filtered image.



**Figure 5.31:** Gaussian and Butterworth low-pass filters.





**Figure 5.32:** High-pass filter used in homomorphic filtering. It is the Butterworth filter damped by a 0.5 coefficients to keep also some low frequencies.



(a)



(b)

**Figure 5.33:** Illustration of homomorphic filtration. (a) Original image. (b) Result of homomorphic filtration. *Courtesy of Tomas Svoboda, Czech Technical University, Prague.*

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# Overview of DTFT properties

We have already discussed (and made use of)

- Periodicity  $X(e^{j\omega}) = X(e^{j(\omega+2\pi)})$
- Linearity

We can group the other properties into meaningful categories:

- 1. properties related to signal symmetry
- 2. properties related to transformations of the independent variable (time domain and frequency domain)
- 3. properties related to time and frequency differentiation
- 4. properties related to convolution
- 5. property related to the energy of the signal (Parseval)
- Our focus will be on understanding the properties and on knowing how to use them, rather than on their mathematical proof

Table 2.1 SYMMETRY PROPERTIES OF THE FOURIER TRANSFORM

**TABLE 2.1** SYMMETRY PROPERTIES OF THE FOURIER TRANSFORM

Sequence $x[n]$	Fourier Transform $X(e^{j\omega})$
1. $x^*[n]$	$X^*(e^{-j\omega})$
2. $x^*[-n]$	$X^*(e^{j\omega})$
3. $\mathcal{R}e\{x[n]\}$	$X_e(e^{j\omega})$ (conjugate-symmetric part of $X(e^{j\omega})$ )
4. $j\mathcal{I}m\{x[n]\}$	$X_o(e^{j\omega})$ (conjugate-antisymmetric part of $X(e^{j\omega})$ )
5. $x_e[n]$ (conjugate-symmetric part of $x[n]$ )	$X_R(e^{j\omega}) = \mathcal{R}e\{X(e^{j\omega})\}$
6. $x_o[n]$ (conjugate-antisymmetric part of $x[n]$ )	$jX_I(e^{j\omega}) = j\mathcal{I}m\{X(e^{j\omega})\}$
<i>The following properties apply only when <math>x[n]</math> is real:</i>	
7. Any real $x[n]$	$X(e^{j\omega}) = X^*(e^{-j\omega})$ (Fourier transform is conjugate symmetric)
8. Any real $x[n]$	$X_R(e^{j\omega}) = X_R(e^{-j\omega})$ (real part is even)
9. Any real $x[n]$	$X_I(e^{j\omega}) = -X_I(e^{-j\omega})$ (imaginary part is odd)
10. Any real $x[n]$	$ X(e^{j\omega})  =  X(e^{-j\omega}) $ (magnitude is even)
11. Any real $x[n]$	$\angle X(e^{j\omega}) = -\angle X(e^{-j\omega})$ (phase is odd)
12. $x_e[n]$ (even part of $x[n]$ )	$X_R(e^{j\omega})$
13. $x_o[n]$ (odd part of $x[n]$ )	$jX_I(e^{j\omega})$

Table 2.2 FOURIER TRANSFORM THEOREMS

**TABLE 2.2** FOURIER TRANSFORM THEOREMS

Sequence	Fourier Transform
$x[n]$	$X(e^{j\omega})$
$y[n]$	$Y(e^{j\omega})$
1. $ax[n] + by[n]$	$aX(e^{j\omega}) + bY(e^{j\omega})$
2. $x[n - n_d]$ ( $n_d$ an integer)	$e^{-j\omega n_d} X(e^{j\omega})$
3. $e^{j\omega_0 n} x[n]$	$X(e^{j(\omega - \omega_0)})$
4. $x[-n]$	$X(e^{-j\omega})$ $X^*(e^{j\omega})$ if $x[n]$ real.
5. $nx[n]$	$j \frac{dX(e^{j\omega})}{d\omega}$
6. $x[n] * y[n]$	$X(e^{j\omega})Y(e^{j\omega})$
7. $x[n]y[n]$	$\frac{1}{2\pi} \int_{-\pi}^{\pi} X(e^{j\theta})Y(e^{j(\omega - \theta)})d\theta$
Parseval's theorem:	
8. $\sum_{n=-\infty}^{\infty}  x[n] ^2 = \frac{1}{2\pi} \int_{-\pi}^{\pi}  X(e^{j\omega}) ^2 d\omega$	
9. $\sum_{n=-\infty}^{\infty} x[n]y^*[n] = \frac{1}{2\pi} \int_{-\pi}^{\pi} X(e^{j\omega})Y^*(e^{j\omega})d\omega$	

Table 2.3 FOURIER TRANSFORM PAIRS

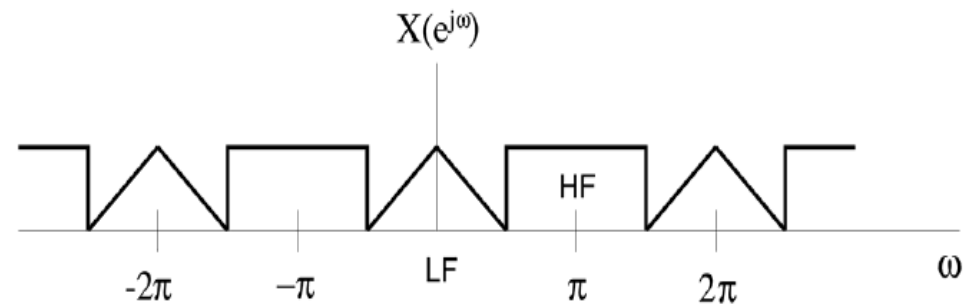
**TABLE 2.3** FOURIER TRANSFORM PAIRS

Sequence	Fourier Transform
1. $\delta[n]$	1
2. $\delta[n - n_0]$	$e^{-j\omega n_0}$
3. 1 $(-\infty < n < \infty)$	$\sum_{k=-\infty}^{\infty} 2\pi\delta(\omega + 2\pi k)$
4. $a^n u[n]$ $( a  < 1)$	$\frac{1}{1 - ae^{-j\omega}}$
5. $u[n]$	$\frac{1}{1 - e^{-j\omega}} + \sum_{k=-\infty}^{\infty} \pi\delta(\omega + 2\pi k)$
6. $(n + 1)a^n u[n]$ $( a  < 1)$	$\frac{1}{(1 - ae^{-j\omega})^2}$
7. $\frac{r^n \sin \omega_p(n + 1)}{\sin \omega_p} u[n]$ $( r  < 1)$	$\frac{1}{1 - 2r \cos \omega_p e^{-j\omega} + r^2 e^{-j2\omega}}$
8. $\frac{\sin \omega_c n}{\pi n}$	$X(e^{j\omega}) = \begin{cases} 1, &  \omega  < \omega_c, \\ 0, & \omega_c <  \omega  \leq \pi \end{cases}$
9. $x[n] = \begin{cases} 1, & 0 \leq n \leq M \\ 0, & \text{otherwise} \end{cases}$	$\frac{\sin[\omega(M + 1)/2]}{\sin(\omega/2)} e^{-j\omega M/2}$
10. $e^{j\omega_0 n}$	$\sum_{k=-\infty}^{\infty} 2\pi\delta(\omega - \omega_0 + 2\pi k)$
11. $\cos(\omega_0 n + \phi)$	$\sum_{k=-\infty}^{\infty} [\pi e^{j\phi} \delta(\omega - \omega_0 + 2\pi k) + \pi e^{-j\phi} \delta(\omega + \omega_0 + 2\pi k)]$

# Frequency shifting: discussion

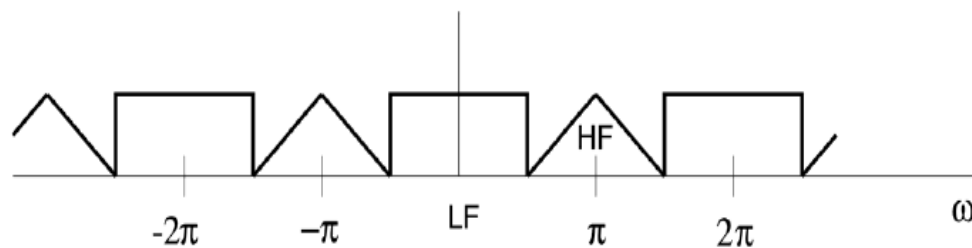
Frequency shifting has important implications because of DTFT periodicity

$$e^{j\omega_0 n} x[n] \longleftrightarrow X(e^{j(\omega - \omega_0)})$$



$$\omega_0 = \pi, y[n] = e^{j\pi n} x[n] = (-1)^n x[n]$$

$$Y(e^{j\omega}) = X(e^{j(\omega - \pi)})$$



# Example 1 (using symmetry properties)

- The following facts are known about a signal  $x[n]$  :

$x[n]$  is real

$x[n] = 0$  for  $n > 0$

$x[0] > 0$

$$\text{Im}\left\{X\left(e^{j\omega}\right)\right\} = \sin \omega - \sin 2\omega$$

$$\frac{1}{2\pi} \int_{-\pi}^{\pi} \left|X\left(e^{j\omega}\right)\right|^2 = 3$$

Determine  $x[n]$



## Example 2 (using the frequency differentiation property)

- Consider the DFT of  $x[n]=a^n u[n]$ , where  $0 < a < 1$
- Show that

$$(n + 1)a^n \xleftrightarrow{DTFT} X(e^{j\omega}) = \frac{1}{(1 - ae^{-j\omega})^2}$$

## Example 3 (using the convolution property)

Consider a discrete LTI system with

$$h[n] = \left(\frac{1}{2}\right)^n u[n]$$

Determine its response to

$$x[n] = \left(\frac{3}{4}\right)^n u[n]$$