Chapter 4 DISCRETE-TIME SYSTEMS 4.5 Introduction to Time-Domain Analysis

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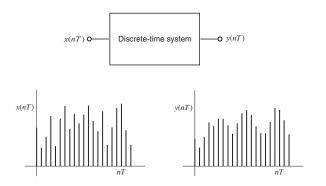
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Introduction

Time-domain analysis is the process of finding the response of a system, y(nT), to a given excitation, x(nT).



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Time-Domain Analysis

Three different methods are available for the time-domain analysis of discrete-time systems:

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Time-Domain Analysis

Three different methods are available for the time-domain analysis of discrete-time systems:

- Induction method
- State-space method
- z transform method

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The *induction method* involves solving the difference equation using induction.

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- The *induction method* involves solving the difference equation using induction.
- The method is somewhat primitive and inefficient.
- However, it is an intuitive method that demonstrates the mode of operation of a discrete-time system.
- It is useful as an introduction to time-domain analysis but it tends to become quite complicated in higher-order discrete-time systems.

State-Space Method

• The *state-space method* entails the manipulation of matrices.

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State-Space Method

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- It is quite useful in applications where routines for the manipulation of matrices are available, e.g., in MATLAB.

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State-Space Method

- The state-space method entails the manipulation of matrices.
- It is quite useful in applications where routines for the manipulation of matrices are available, e.g., in MATLAB.
- It is applicable to time-dependent systems.

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Z Transform Method

The z transform method is the most efficient and most frequently used method among the available methods.

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Z Transform Method

The z transform method is the most efficient and most frequently used method among the available methods.

 Its main disadvantage is that it cannot be applied to time-dependent or nonlinear systems.

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Z Transform Method

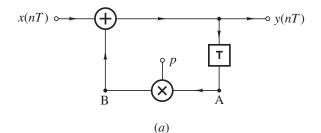
- The z transform method is the most efficient and most frequently used method among the available methods.
- Its main disadvantage is that it cannot be applied to time-dependent or nonlinear systems.
- The details of the method can be found in Chap. 5.

Induction Method Cont'd

- The induction method for time-domain analysis can be illustrated by finding the impulse, unit-step, and sinusoidal response of a simple recursive system.
 - As will be shown, all that is necessary is simple algebra.



Find the impulse response of the recursive system



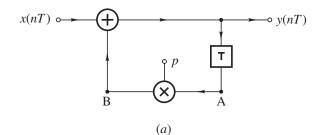
assuming an initially relaxed system.

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Solution The difference equation is

$$y(nT) = x(nT) + py(nT - T)$$

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Example Cont'd

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$$y(nT) = x(nT) + py(nT - T)$$

If $x(nT) = \delta(nT)$, we have

$$y(nT) = \delta(nT) + py(nT - T)$$

For an initially relaxed system, y(nT) = 0 for n < 0 and hence we have

$$y(0) = \delta(0) + py(-T) = 1 + 0 = 1$$

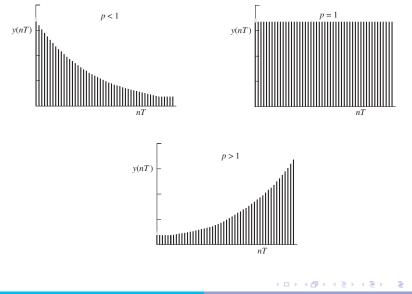
$$y(T) = \delta(T) + py(0) = 0 + p \times 1 = p$$

$$y(2T) = \delta(2T) + py(T) = 0 + p \cdot p = p^{2}$$

$$\vdots$$

$$y(nT) = u(nT)p^{n} \quad \blacksquare$$

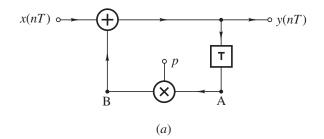
The unit-step u(nT) is added to ensure that y(nT) = 0 for n < 0.



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Example

Assuming that the system shown is initially relaxed, find the unit-step response:



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Example Cont'd

Solution The difference equation is

$$y(nT) = x(nT) + py(nT - T)$$

If x(nT) = u(nT), we have

$$y(nT) = u(nT) + py(nT - T)$$

For an initially relaxed system, y(nT) = 0 for n < 0 and hence we have

$$y(0) = u(0) + py(-T) = 1 + 0 = 1$$

$$y(T) = u(T) + py(0) = 1 + p$$

$$y(2T) = u(2T) + py(T) = 1 + p + p^{2}$$

$$\vdots$$

$$y(nT) = u(nT) \sum_{k=0}^{n} p^{k}$$

 $y(nT) = u(nT)\sum_{k=0}^{n} p^{k}$

We can write

$$y(nT) = u(nT)(1 + p + p^{2} + \dots + p^{n})$$
(A)
$$py(nT) = u(nT)(p + p^{2} + \dots + p^{n} + p^{(n+1)})$$
(B)

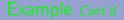
Subtracting Eq. (B) from Eq. (A), we get

$$y(nT) - py(nT) = u(nT)(1 - p^{(n+1)})$$

or

$$y(nT) = u(nT)\frac{1-p^{(n+1)}}{1-p}$$

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Therefore, there are three cases to consider:

(i) p < 1(ii) p = 1(iii) p > 1

(i) For p < 1, the steady-state response is obtained by evaluating y(nT) for n→∞, i.e.,

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$$\lim_{n\to\infty}y(nT)=\frac{1}{1-p}$$

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Digital Signal Processing – Sec. 4.5

(i) For p < 1, the steady-state response is obtained by evaluating y(nT) for n→∞, i.e.,

$$\lim_{n\to\infty}y(nT)=\frac{1}{1-p}$$

(ii) For p = 1, using l'Hôpital's rule we get

$$y(nT) = \lim_{p \to 1} \frac{d(1 - p^{(n+1)})/dp}{d(1 - p)/dp} = n + 1$$

Hence

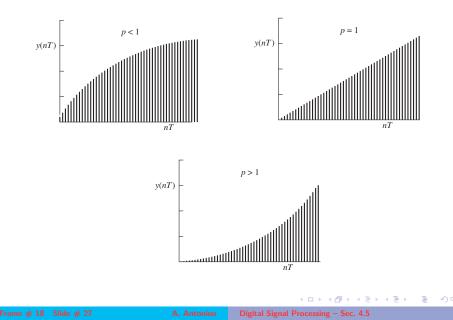
$$\lim_{n\to\infty}y(nT)\to\infty$$

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(iii) For
$$p > 1$$

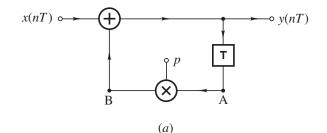
$$\lim_{n\to\infty}y(nT)\approx\frac{p^n}{p-1}\to\infty\quad \bullet$$

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Example

Assuming zero initial conditions, find the response of the recursive system



to the sinusoidal excitation

 $x(nT) = u(nT)\sin\omega nT$

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Example Cont'd

Solution As before, the difference equation is

$$y(nT) = x(nT) + py(nT - T)$$

If $x(nT) = u(nT) \sin \omega nT$, we have

$$y(nT) = \mathcal{R}x(nT) = u(nT)\sin\omega nT + py(nT - T)$$

The system is linear and so

$$y(nT) = \mathcal{R}[u(nT)\sin\omega nT] = \mathcal{R}\left[u(nT)\frac{1}{2j}(e^{j\omega nT} - e^{-j\omega nT})\right]$$
$$= \frac{1}{2j}\left[\mathcal{R}u(nT)e^{j\omega nT} - \mathcal{R}u(nT)e^{-j\omega nT}\right]$$
$$= \frac{1}{2j}[y_1(nT) - y_2(nT)]$$

where

$$y_1(nT) = \mathcal{R}u(nT)e^{j\omega nT}$$
 and $y_2(nT) = \mathcal{R}u(nT)e^{-j\omega nT}$

$$y(nT) = x(nT) + py(nT - T)$$

 $y_1(nT) = \mathcal{R}u(nT)e^{j\omega nT}$ and $y_2(nT) = \mathcal{R}u(nT)e^{-j\omega nT}$

The partial response $y_1(nT)$ can be obtained as

$$y_1(nT) = \mathcal{R}\left[u(nT)e^{j\omega nT}\right] = u(nT)e^{j\omega nT} + py_1(nT - T)$$

Hence

$$y_{1}(0) = u(0)e^{0} + py_{1}(-T) = 1$$

$$y_{1}(T) = e^{j\omega T} + py_{1}(0) = e^{j\omega T} + p$$

$$y_{1}(2T) = e^{j2\omega T} + py_{1}(T) = e^{j2\omega T} + pe^{j\omega T} + p^{2}$$

$$\vdots$$

$$y_{1}(nT) = u(nT)(e^{j\omega nT} + pe^{j\omega(n-1)T} + \dots + p^{(n-1)}e^{j\omega T} + p^{n})$$

$$= u(nT)e^{j\omega nT}(1 + pe^{-j\omega T} + \dots + p^{n}e^{(-jn\omega T)})$$

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$$y_1(nT) = u(nT)e^{j\omega nT}(1 + pe^{-j\omega T} + \dots + p^n e^{(-jn\omega T)})$$
$$= u(nT)e^{j\omega nT} \sum_{k=0}^n p^k e^{(-jk\omega nT)}$$

This is a geometric series in powers of $pe^{(-j\omega nT)}$ and its sum can be obtained as

$$y_1(nT) = u(nT)\frac{e^{j\omega nT} - p^{(n+1)}e^{-j\omega T}}{1 - pe^{-j\omega T}}$$

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$$y_1(nT) = u(nT)\frac{e^{j\omega nT} - p^{(n+1)}e^{-j\omega T}}{1 - pe^{-j\omega T}} = \frac{e^{j\omega T}}{e^{j\omega T} - p} \times (e^{j\omega nT} - p^{(n+1)}e^{-j\omega T})$$

Now consider the function

$$H(e^{j\omega T}) = \frac{e^{j\omega T}}{e^{j\omega T} - p} = \frac{\cos \omega T + j \sin \omega T}{\cos \omega T + j \sin \omega T - p}$$

and let

$$H(e^{j\omega T}) = M(\omega)e^{j\theta(\omega)}$$

where $M(\omega) = |H(e^{j\omega T})| = \frac{1}{\sqrt{1 + p^2 - 2p \cos \omega T}}$ and $\theta(\omega) = \arg H(e^{j\omega T}) = \omega T - \tan^{-1} \frac{\sin \omega T}{\cos \omega T - p}$

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$y_1(nT) = rac{e^{j\omega T}}{e^{j\omega T} - p} imes (e^{j\omega nT} - p^{(n+1)}e^{-j\omega T})$

and

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and $\theta(\omega) = \arg H(e^{j\omega T}) = \omega T - \tan^{-1} \frac{\sin\omega T}{\cos\omega T - p}$

By using these relations, $y_1(\omega T)$ can be expressed as $y_1(nT) = u(nT)M(\omega)(e^{j[\theta(\omega)+\omega nT]} - p^{(n+1)}e^{j[\theta(\omega)-\omega T]})$

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$$y_1(nT) = u(nT)M(\omega)(e^{j[\theta(\omega)+\omega nT]} - p^{(n+1)}e^{j[\theta(\omega)-\omega T]})$$

By replacing ω by $-\omega$ in $y_1(nT)$, we get

$$y_2(nT) = u(nT)M(\omega)(e^{j[\theta(-\omega)-\omega nT]} - p^{(n+1)}e^{j[\theta(-\omega)+\omega T]})$$

By noting that $M(\omega)$ is an even function and $\theta(\omega)$ an odd function of ω , i.e.,

$$M(-\omega) = M(\omega)$$
 and $\theta(-\omega) = -\theta(\omega)$

we can readily show that

$$y(nT) = \frac{1}{2j} [y_1(nT) - y_2(nT)]$$

= $u(nT)M(\omega) \sin[\omega nT + \theta(\omega)]$
 $-u(nT)M(\omega)p^{(n+1)} \sin[\theta(\omega) - \omega T]$

$$y(nT) = u(nT)M(\omega)\sin[\omega nT + \theta(\omega)] -u(nT)M(\omega)p^{(n+1)}\sin[\theta(\omega) - \omega T]$$

As can be seen, the sinusoidal response of the system consists of two components.

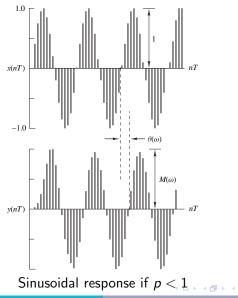
If p < 1, the second term represents a transient component that reduces to zero as $n \to \infty$. Therefore,

$$\tilde{y}(nT) = \lim_{n \to \infty} y(nT) = M(\omega) \sin[\omega nT + \theta(\omega)]$$

If p = 1, the transient component is a constant. If p > 1 the transient component tends to infinity as $n \to \infty$ i.e.,

$$\tilde{y}(nT) = \lim_{n \to \infty} y(nT) \to \infty$$

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Digital Signal Processing - Sec. 4.5

Sinusoidal Response

The time-domain analysis has shown that the response of a first-order recursive system to a sinusoidal excitation of unity amplitude and zero phase angle, i.e.,

$$x(nT) = \sin(\omega nT)$$

is a sinusoid of amplitude $M(\omega)$ and angle $\theta(\omega)$, i.e.,

$$x(nT) = M(\omega)\sin[\omega nT + \theta(\omega)]$$

provided that the transient component decays to zero.

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It turns out that this is a general property of recursive as well as nonrecursive systems in general.

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provided that the transient component decays to zero.

- It turns out that this is a general property of recursive as well as nonrecursive systems in general.
- The transient response of a discrete-time system will decay to zero only if the system is stable (see Sec. 4.7).

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This slide concludes the presentation. Thank you for your attention.